

**Low cost 6-bit monolithic
D-A converter**

**ZN436E
ZN436J**

FEATURES

- 6-bit accuracy
- ZN436E commercial temperature range
0 to +70°C
- ZN436J military temperature range
-55 to +125°C
- TTL and 5V CMOS compatible
- Single +5V supply
- Settling time 1μsec. typical
- Designed for low-cost applications

DESCRIPTION

The ZN436 is a monolithic 6-bit D-A converter containing an R-2R ladder network of diffused resistors with precision bipolar switches.

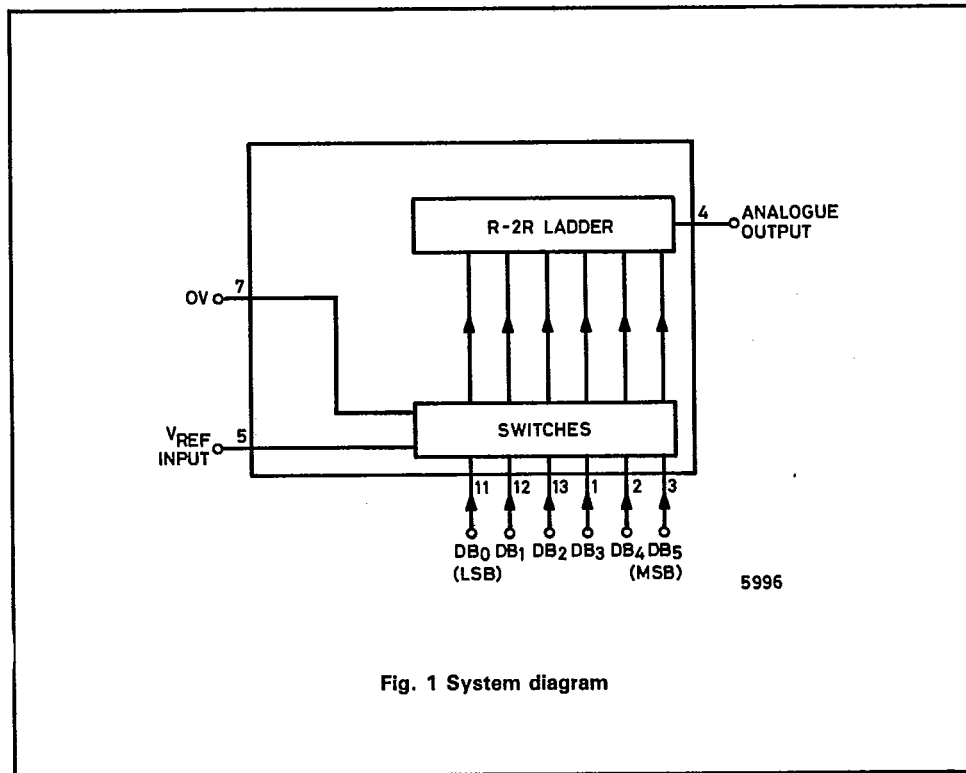


Fig. 1 System diagram

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INTRODUCTION

The ZN436 is an 6-bit D-A converter. It contains an advanced design of R-2R ladder network and an array of precision bipolar switches on a single monolithic chip.

full 6-bit accuracy using normal diffused resistors.

The converter is of the voltage switching type and uses an R-2R resistor ladder network as shown in Fig. 2.

The special design of ladder network results in

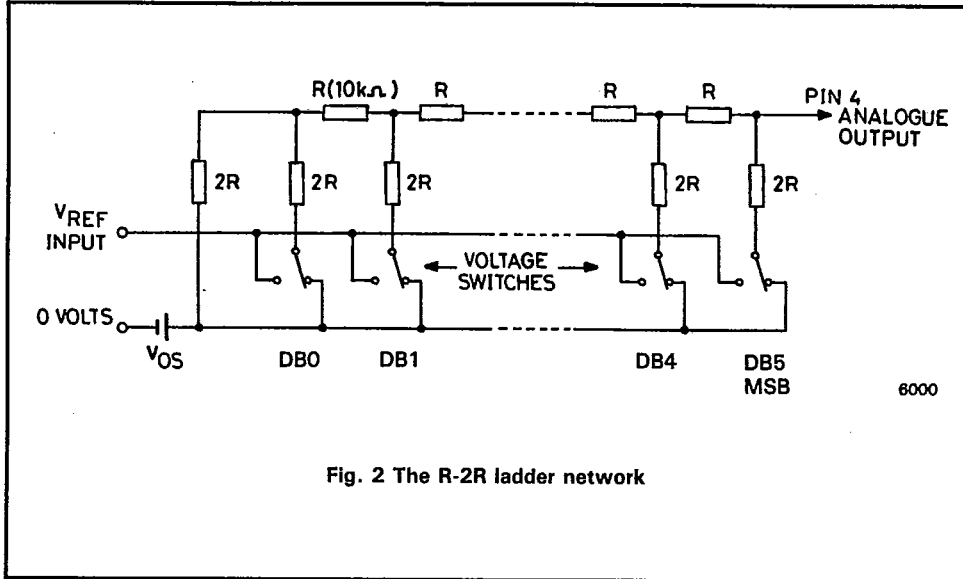


Fig. 2 The R-2R ladder network

Each 2R element is connected either to 0V or V_{REF} by transistor switches specially designed for low offset voltage (typically 1mV).

which should have a slope resistance less than 2Ω

Binary weighted voltages are produced at the output of the R-2R ladder, the value depending on the digital number applied to the bit inputs.

Suggested external reference sources are the ZN404 or one of the ZN458 range. Each ZN404 is capable of supplying up to five ZN436 circuits and this is increased to ten for the ZN458 range.

An external fixed or varying reference is required

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ORDERING INFORMATION

Operating temperature	6-Bit accuracy	Package
0 to +70°C	ZN436E	Plastic
-55 to +125°C	ZN436J	Ceramic

ABSOLUTE MAXIMUM RATINGS

Supply voltage V_{CC} +7.0V
 Max. voltage, logic and V_{REF} inputs +5.5V
 Storage temperature range -55 to +125°C

CHARACTERISTICS (at $T_{amb} = 25^\circ C$ and $V_{CC} = +5V$ unless otherwise specified)

Parameter	Symbol	Min.	Typ.	Max.	Units	Conditions
Accuracy (useful resolution)						V_{REF} input = 2.0 to 3.0V
ZN436J		6	-	-	bits	
ZN436E		6	-	-	bits	
Non-linearity		-	-	± 0.5	L.S.B.	Note 1
Differential non-linearity		-	± 0.5	-	L.S.B.	Note 2
Settling time to 0.5 L.S.B.		-	1.0	-	μs	1 L.S.B. step
Settling time to 0.5 L.S.B.		-	2.0	-	μs	All bits ON to OFF or OFF to ON
Offset voltage						All bits OFF note 1
ZN436J	V_{OS}	-	5.0	8.0	mV	
ZN436E		-	3.0	5.0	mV	
V_{OS} temperature coefficient		-	5.0	-	$\mu V/^\circ C$	
Full-scale output		2.510	2.520	2.530	V	All bits ON Ext. $V_{REF} = 2.560V$
Full-scale temp. coefficient		-	3.0	-	ppm/ $^\circ C$	Ext. $V_{REF} = 2.560V$
Non-linearity temp. coefficient		-	7.5	-	ppm/ $^\circ C$	Relative to F.S.R.

Notes:

1. The ZN436J differs from the ZN436E in the following respects:
 - (a) For the ZN436J, the maximum linearity error may increase to $\pm 1LSB$ over the temperature ranges -55 to 0°C and +70 to +125°C.
 - (b) Offset voltage. The difference is due to package lead resistance. This offset will normally be removed by the setting up procedure, and because the offset temperature coefficient is low, the specified accuracy will be maintained.
2. Monotonic over the full operating temperature range.

ZN436E ZN436J**CHARACTERISTICS (cont.)**

Parameter	Symbol	Min.	Typ.	Max.	Units	Conditions
Analogue output resistance	R_o	—	10	—	$k\Omega$	
External reference voltage		0	—	3.0	V	
Supply voltage	V_{CC}	4.5	—	5.5	V	
Supply current	I_S	—	5	9	mA	
High level input voltage	V_{IH}	2.0	—	—	V	
Low level input voltage	V_{IL}	—	—	0.7	V	
High level input current	I_{IH}	—	—	10	μA	$V_{CC} = \text{max.}, V_I = 2.4V$
		—	—	100	μA	$V_{CC} = \text{max.}, V_I = 5.5V$
Low level input current	I_{IL}	—	—	-0.18	mA	$V_{CC} = \text{max.}, V_I = 0.3V$

APPLICATIONS**1. 6-bit D-A converter**

The ZN436 gives an analogue voltage output directly from pin 4 therefore the usual current to voltage converting amplifier is not required. The output voltage drift, due to the temperature coefficient of the analogue output resistance R_o , will be less than 0.004% per °C (or 1LSB/100°C) if R_I is chosen to be $\geq 650k\Omega$.

In order to remove the offset voltage and to calibrate the converter a buffer amplifier is necessary. Fig. 3 shows a typical scheme. To minimise temperature drift in this and similar

applications the source resistance to the inverting input of the operational amplifier should be approximately $6k\Omega$. The calibration procedure is as follows:

- i. Set all bits to OFF (low) and adjust R_2 until $V_{OUT} = 0.000V$.
- ii. Set all bits to ON (high) and adjust R_1 until $V_{OUT} = \text{Nominal full-scale reading} - 1\text{LSB}$.
- iii. Repeat i. and ii.

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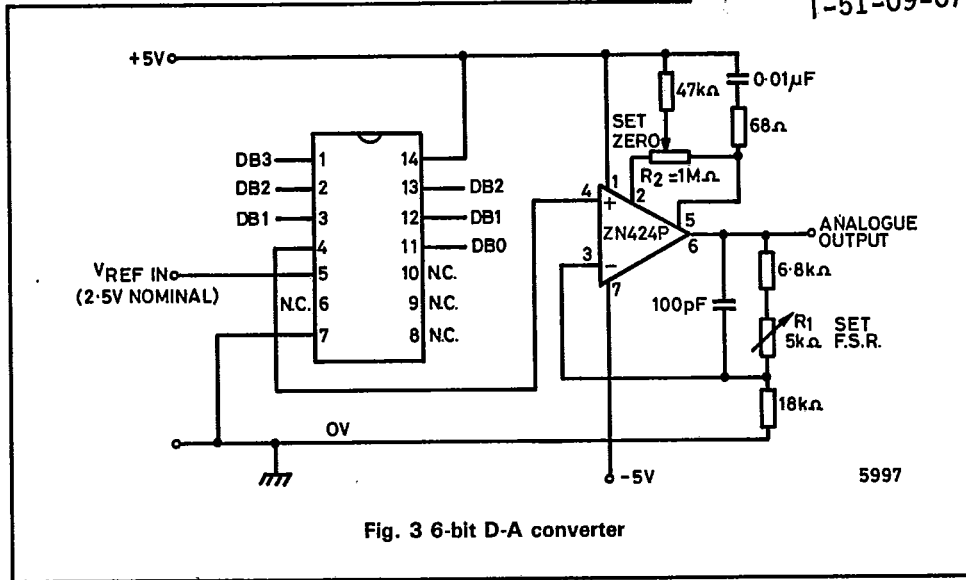


Fig. 3 6-bit D-A converter

Alternative output buffer using the ZLD741
The following circuit, employing the ZLD741

operational amplifier, may be used as the out-
put buffer (Fig. 3).

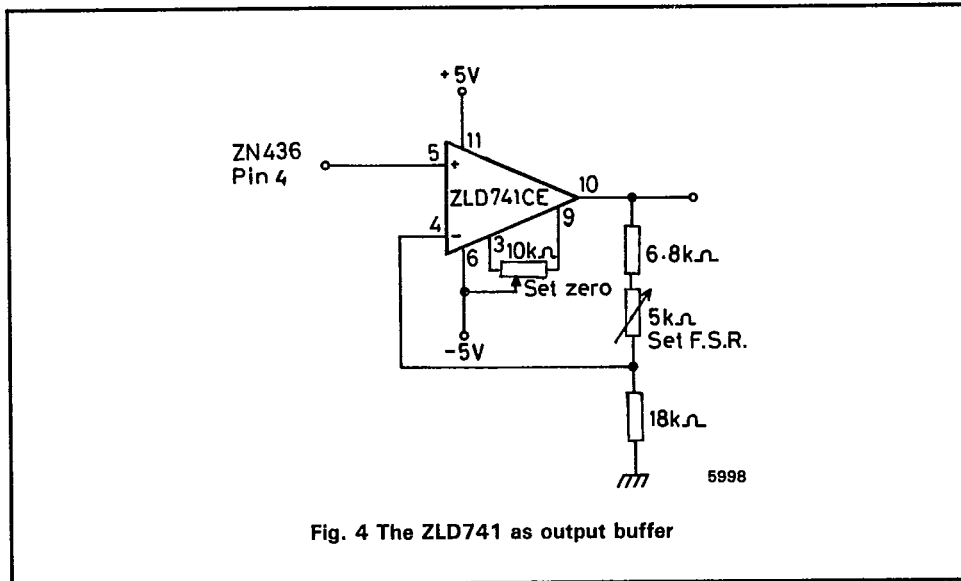


Fig. 4 The ZLD741 as output buffer

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PIN CONNECTIONS

